Two Methods of Obtaining a Minimal Upper Estimate for the Error Probability of the Restoring Formal Neuron

A. I. Prangishvili, O. M. Namicheishvili, M. A. Gogiashvili

Abstract— It is shown that a minimal upper estimate for the error probability of the formal neuron, when the latter is used as a restoring (decision) element, can be obtained by the Laplace transform of the convolution of functions as well as by means of the generating function of the factorial moment of the sum of independent random variables. It is proved that in both cases the obtained minimal upper estimates are absolutely identical.

Index Terms—generating function, probability of signal restoration error, restoring neuron, upper estimate.

I. INTRODUCTION

Let us consider the formal neuron, to the inputs of which different versions $X_1, X_2, ..., X_n, X_{n+1}$ of one and the same random binary signal X arrive via the binary channels $B_1, B_2, ..., B_n, B_{n+1}$ with different error probabilities $q_i (i = \overline{1, n+1})$, and the neuron must restore the correct input signal X or, in other words, make a decision Y using the versions $X_1, X_2, ..., X_n, X_{n+1}$. When the binary signal X arrives at the inputs of the restoring element via the channels of equal reliability, the decision-making, in which some value prevails among the signal versions, i.e. the decision-making by the majority principle, was for the first time described by J. von Neumann [1], and later V. I. Varshavski [2] generalized this principle to redundant analog systems.

In the case of input channels with different reliabilities, adaptation of the formal neuron is needed in order to restore the correct signal. Adaptation is interpreted as the control process of weights $a_i (i = \overline{1, n+1})$ of the neuron inputs, which makes these weights match the current probabilities $q_i (i = \overline{1, n+1})$ of the input channels. The purpose of this control is to make inputs of high reliability to exert more influence on decision-making (i.e. on the restoration of the correct signal) as compared with inputs of low reliability. Restoration is carried out by vote-weighting by the relation

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$$Y = \operatorname{sgn}\left(\sum_{i=1}^{n+1} a_i X_i\right) = \operatorname{sgn} Z, \qquad (1)$$

where

$$Z = \sum_{i=1}^{n+1} a_i X_i \ .$$
 (2)

Both the input signal X and its versions X_i (i = 1, n) are considered as binary random variables coded by the logical values (+1) and (-1). It is formally assumed that the threshold Θ of the restoring neuron is introduced into consideration by means of the identity $\Theta \equiv a_{n+1}$, where $(-\infty < a_{n+1} < \infty)$ and the signal $X_{n+1} \equiv -1$. The main point of this formalism is that the signal $X_{n+1} \equiv -1$ is dumped from some imaginary binary input B_{n+1} for any value of the input signal X, whereas the value q_{n+1} is the a priori probability of occurrence of the signal X = +1 or, which is the same, the error probability of the channel B_{n+1} . Quite a vast literature [3]-[7] is dedicated to threshold logic which takes into consideration the varying reliability of channels, but in this paper we express our viewpoint in the spirit of the ideas of W. Pierce [8].

Let us further assume that

$$\operatorname{sgn} Z = \begin{cases} -1 & \text{if } Z < 0 \\ +1 & \text{if } Z \ge 0 \end{cases}.$$
(3)

When Z = 0, the solution Y at the output of the restoring formal neuron has the form +1 according to (3). The probability that the restored value Y of the signal X is not correct is expressed by the formula

$$Q = \operatorname{Prob}\left\{Y \neq X\right\} = \operatorname{Prob}\left\{\eta < 0\right\}.$$
(4)

Here $\eta = XZ$ is a discrete random variable with probability distribution density f(v). This variable is the sum of independent discrete variables $\eta_i = a_i XX_i$, and the function $f_i(v_i)$ describes the probability distribution density of individual summands η_i . For the realizations of random variables η and η_i we introduce the symbols v and v_i , respectively.

It is easy to observe that the variable η_i takes the values $+a_i$ and $-a_i$ with probabilities $1-q_i$ and q_i , respectively. Therefore, if we use the Dirac delta function $\delta(t)$, then the probability density $f_i(v_i)$ can be represented as follows

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$$\begin{cases} f_i(v_i) = (1 - q_i)\delta(v_i - a_i) + q_i\delta(v_i + a_i) \\ v_i = +a_i, -a_i \\ i = \overline{1, n+1} \end{cases} .$$
 (5)

Such formalism is completely justified and frequently used due to the following two properties of the delta-function

$$\delta(t) \ge 0, \,\forall t \in \mathbb{R}$$

$$\int_{-\infty}^{+\infty} \delta(t) dt = 1$$

However $f_i(v_i)$ can also be represented as

$$\left. \begin{cases} f_i(v_i) = q_i^{(a_i - v_i)/2a_i} \cdot (1 - q_i)^{(a_i + v_i)/2a_i} \\ v_i = +a_i, -a_i \\ i = \overline{1, n+1} \end{cases} \right\}.$$
(6)

The random variable η is the sum of independent discrete random variables η_i . Its distribution density f(v) can be defined in the form of convolution of probability distribution densities of summands $f_i(v_i)$:

$$f(v) = \sum_{i=1}^{n+1} f_i(v_i) , \qquad (7)$$

where * (superposition of the addition and multiplication signs) is the convolution symbol.

It is obvious that in view of formula (7) the error probability at the decision element output can be written in two equivalent forms

$$Q = \operatorname{Prob}(v < 0) = \int_{-\infty}^{0} f(v) dv = \int_{-\infty}^{0} \int_{-\infty}^{n+1} f_i(v_i) dv =$$

$$= \int_{-\infty}^{0} \int_{i=1}^{n+1} \left[(1 - q_i) \delta(v_i - a_i) + q_i \delta(v_i + a_i) \right] dv$$
(8)

and

$$Q = \sum_{v < 0} f(v) = \sum_{v < 0}^{n+1} f_i(v_i) =$$

$$= \sum_{v < 0}^{n+1} \sum_{i=1}^{n+1} \left[q_i^{(a_i - v_i)/2a_i} \cdot (1 - q_i)^{(a_i + v_i)/2a_i} \right],$$
(9)

where the probability distribution density $f_i(v_i)$ is defined by (5) in the first case and by (6) in the second case. Integration or summation in both cases is carried out continuously or discretely over all negative values of the variable v. Formulas (8) and (9) give an exact value of the error probability of restoration of a binary signal by the formal neuron.

Note that for practical calculations, formula (9) can be written in a more convenient form. Indeed, the complete number of discrete values of the variable v is 2^{n+1} since

$$v = \widetilde{a_1} + \widetilde{a_2} + \dots + \widetilde{a_n} + \widetilde{a_{n+1}},$$

where $\widetilde{a_i}$ is equal either to $+(a_i)$ or to $-(a_i)$, whereas the proper sign of the weight a_i is meant to be within the round brackets.

By formula (9), to each discrete value of the sum v there corresponds the term Q_j ($j = \overline{1, 2^{n+1}}$) which is the product of (n+1) co-factors of the form q_k or $(1-q_k)$.

In particular

$$Q_{j} \equiv f(v) = \underset{i=1}^{n+1} f_{i}(v_{i}) \equiv \widetilde{q_{1}} \cdot \widetilde{q_{2}} \cdot \cdots \cdot \widetilde{q_{n}} \cdot \widetilde{q_{n+1}}$$

$$j = \overline{1, 2^{n+1}}$$

where

$$\widetilde{q_k} = \begin{cases} q_k & \text{if } v_k = -(a_k) \\ 1 - q_k & \text{if } v_k = +(a_k) \end{cases}$$

for any $k\left(k=\overline{1,n+1}\right)$.

Thus formula (9) can also be written in the form

$$Q = \sum_{\nu < 0} Q_j = \sum_{\nu < 0} \widetilde{q_1} \cdot \widetilde{q_2} \cdot \dots \cdot \widetilde{q_n} \cdot \widetilde{q_{n+1}}, \qquad (10)$$

which is more adequate for cognitive perception and practical realization.

II. FINDING A MINIMAL UPPER ESTIMATE BY THE FIRST METHOD

From the expression

$$Q = \operatorname{Prob}(\eta < 0) = \int_{-\infty}^{0} f(v) dv$$

it follows that for a real positive number s (s > 0)

$$Q \leq \int_{-\infty}^{0} e^{-sv} f(v) dv \leq \int_{-\infty}^{\infty} e^{-sv} f(v) dv.$$

But the left-hand part of this inequality is the Laplace transform of the function f(v)

$$\mathfrak{L}[f(v)] = \int_{-\infty}^{\infty} e^{-sv} f(v) dv,$$

where \mathfrak{L} is the Laplace transform operator.

$$Q \le \mathfrak{L}[f(v)]. \tag{11}$$

The random value η with realizations v is the sum of independent random variables η_i having realizations v_i . In that case, as is known, the Laplace transform for the convolution f(v) of functions $f_i(v_i)$ is equal to the product of Laplace transforms of convoluted functions:

$$\mathfrak{L}[f(\mathbf{v})] = \prod_{i=1}^{n+1} \mathfrak{L}[f_i(\mathbf{v}_i)].$$

The latter implies that

Therefore

$$Q \le \prod_{i=1}^{n+1} \mathfrak{L}[f_i(v_i)].$$
(12)

By expression (5) for functions $f_i(v_i)$ and the Laplace transform definition, we obtain

$$\mathcal{L}[f_i(v_i)] = \int_{-\infty}^{\infty} e^{-sv_i} f_i(v_i) dv_i =$$
$$= \int_{-\infty}^{\infty} e^{-sv_i} \left[(1-q_i) \delta(v_i - a_i) + q_i \delta(v_i + a_i) \right] dv_i$$

Using this expression in formula (12) we have

$$Q \le \prod_{i=1}^{n+1} \int_{-\infty}^{\infty} e^{-sv_i} \left[(1-q_i)\delta(v_i - a_i) + q_i\delta(v_i + a_i) \right] dv_i.$$
(13)

Here we should make use of one more property of the Dirac delta function

$$\int_{-\infty}^{\infty} g(t)\delta(t-t_0)dt = g(t_0).$$

With this property taken into account, from formula (13) we obtain

$$Q \le \prod_{i=1}^{n+1} \left[(1-q_i) e^{-a_i s} + q_i e^{+a_i s} \right].$$
(14)

Here s, as mentioned above, is an arbitrary real positive number. Before we continue simplifying the right-hand part of inequality (14), we have to define a value of s for which expression (14) gives *a minimal upper estimate*.

Passing to the natural logarithm of inequality (14) we come to the expression

$$\ln Q \le \sum_{i=1}^{n+1} \ln \left[(1-q_i) e^{-a_i s} + q_i e^{+a_i s} \right].$$

Let us define here partial derivatives with respect to arguments a_i by using the elementary fact that

$$y' = \frac{dy}{dx} = f'(x) \cdot e^{f(x)}$$

if $y = e^{f(x)}$, and also the fact that $\frac{d}{dx} \ln f(x) = f'(x) \cdot \frac{1}{f(x)}$.

Hence we obtain

$$\frac{\partial \ln Q}{\partial a_i} \le \sum_{i=1}^{n+1} \frac{1}{[(1-q_i)e^{-a_i s} + q_i e^{+a_i s}]} \cdot [sq_i e^{+a_i s} - s(1-q_i)e^{-a_i s}].$$

For the right-hand part of this inequality to be equal to zero, it suffices that the following condition be fulfilled

$$sq_i e^{+a_i s} - s(1-q_i)e^{-a_i s} = 0,$$

whence it follows that

$$e^{+2a_is}=\frac{1-q_i}{q_i},$$

or, which is the same,

$$a_i s = \frac{1}{2} \ln \frac{1 - q_i}{q_i}.$$

If the weights a_i of the neuron inputs are put into correspondence with error probabilities q_i of these inputs by the relations

$$a_i = \ln \frac{1 - q_i}{q_i},\tag{15}$$

then the sought value of s will be

$$s = \frac{1}{2}.$$
 (16)

Using equality (16) in formula (14), we obtain a minimal upper estimate for the error probability Q of the restoring neuron. Indeed, for the right-hand part of expression (14) the following chain of identical transforms is valid:

$$q_{i}e^{\frac{a_{i}}{2}} + (1-q_{i})e^{-\frac{a_{i}}{2}} = 2\sqrt{q_{i}(1-q_{i})} \cdot \frac{q_{i}e^{\frac{a_{i}}{2}} + (1-q_{i})e^{-\frac{a_{i}}{2}}}{2\sqrt{q_{i}(1-q_{i})}} = 2\sqrt{q_{i}(1-q_{i})} \cdot \frac{\sqrt{\frac{q_{i}}{1-q_{i}}} \cdot e^{\frac{a_{i}}{2}} + \sqrt{\frac{1-q_{i}}{q_{i}}} \cdot e^{-\frac{a_{i}}{2}}}{2}.$$

Let us take into account here that

$$\sqrt{\frac{q_i}{1-q_i}} = \exp\left[\ln\left(\sqrt{\frac{q_i}{1-q_i}}\right)\right] = \exp\left[-\frac{1}{2}\ln\frac{1-q_i}{q_i}\right]$$
$$\sqrt{\frac{1-q_i}{q_i}} = \exp\left[\ln\left(\sqrt{\frac{1-q_i}{q_i}}\right)\right] = \exp\left[+\frac{1}{2}\ln\frac{1-q_i}{q_i}\right]$$

Besides, we denote

$$\frac{1}{2}\left(a_i - \ln\frac{1-q_i}{q_i}\right) = \lambda_i.$$

Then we have

$$q_i e^{\frac{a_i}{2}} + (1-q_i) e^{-\frac{a_i}{2}} = 2\sqrt{q_i(1-q_i)} \cdot \frac{e^{\lambda_i} + e^{-\lambda_i}}{2}.$$

The second co-factor in the right-hand part of this expression is the hypebolic cosine of the argument λ_i :

$$\frac{e^{\lambda_i} + e^{-\lambda_i}}{2} = \operatorname{ch} \lambda_i.$$

Therefore

$$q_i e^{\frac{a_i}{2}} + (1-q_i) e^{-\frac{a_i}{2}} = 2\sqrt{q_i(1-q_i)} \cdot \operatorname{ch} \lambda_i =$$
$$= 2\sqrt{q_i(1-q_i)} \cdot \operatorname{ch} \left(\frac{a_i - \ln \frac{1-q_i}{q_i}}{2}\right).$$

Finally, for estimate (14) we can write

$$Q \leq \prod_{i=1}^{n+1} \left\{ 2\sqrt{q_i(1-q_i)} \cdot \operatorname{ch}\left(\frac{a_i - \ln\frac{1-q_i}{q_i}}{2}\right) \right\}$$

For the error probability Q, the right-hand part of the above inequality is the upper estimate Q^+ :

$$Q^{+} = \prod_{i=1}^{n+1} \left[2\sqrt{q_i(1-q_i)} \cdot \operatorname{ch}\left(\frac{a_i - \ln\frac{1-q_i}{q_i}}{2}\right) \right].$$

The minimum Q_{\min}^+ of this upper estimate Q^+ is equal to

$$Q_{\min}^{+} = \prod_{i=1}^{n+1} \left[2\sqrt{q_i(1-q_i)} \right] = 2^{n+1} \cdot \prod_{i=1}^{n+1} \left[\sqrt{q_i(1-q_i)} \right].$$
(17)

It is attained when for the zero argument the hyperbolic cosine attains the minimum equal to 1.

This estimate confirms in a certain sense the advantage of the choice of weights of the restoring neuron in compliance with the error probabilities of input signals according to the following relations

$$\left. \begin{array}{l} a_i = \ln \frac{1 - q_i}{q_i} \\ i = \overline{1, n+1} \end{array} \right\}$$

III. OBTAINING A MINIMAL UPPER ESTIMATE BY THE SECOND METHOD

Simulteously, for the probability Q it is useful to obtain a minimal upper estimate in the closed analytic form by one more new approach.

As is known [9], the generating function $\gamma_{\nu}(S)$ of the factorial moment of the sum η of independent random variables η_i is equal to the product of generating functions $\gamma_{\nu_i}(S)$ of the factorial moments of individual summands, i.e.

$$\gamma_{\nu}(S) = \prod_{i=1}^{n+1} \gamma_{\nu_i}(S), \qquad (18)$$

where

$$\gamma_{\nu}(S) = \mathbf{M} \Big[S^{\eta} \Big] = \sum_{\nu} S^{\nu} \cdot f(\nu) , \qquad (19)$$

$$\gamma_{\nu_i}(S) = \mathbf{M} \left[S^{\eta_i} \right] = \sum_{\nu_i} S^{\nu_i} \cdot f_i(\nu_i) \left\{ i = \overline{1, n+1} \right\}.$$
(20)

Here M is the mathematical expectation symbol and S is an arbitrary complex number for which series (19) and (20) exist on some segment of the real axis containing the point S = 1.

Since in relation (20), summation is carried out on the set of two possible values $+a_i$ and $-a_i$ of the variable v_i , using (6) we have

$$\gamma_{\nu_i}(S) = (1 - q_i)S^{a_i} + q_iS^{-a_i} \\ i = \overline{1, n+1}$$
 (21)

The substitution of (21) into relation (18) gives

$$\gamma_{\nu}(S) = \prod_{i=1}^{n+1} \left[(1-q_i) S^{a_i} + q_i S^{-a_i} \right].$$

When v < 0, the value S^{v} satisfies the condition

$$S^{\nu} = \frac{1}{S^{(\nu)}} > 1,$$

if of course

Let us assume that inequality (22) is fulfilled. Then the following relation is valid:

0 < *S* < 1.

$$Q = \sum_{v<0} f(v) < \sum_{v<0} S^{v} \cdot f(v).$$

Since every summand $S^{\nu} \cdot f(\nu)$ is non-negative, we have the inequality

 $\sum_{v<0} S^{v} f(v) \leq \sum_{v} S^{v} \cdot f(v) \,.$

Therefore

$$Q < \gamma_{\nu}(S) . \tag{23}$$

The right-hand part of this expression can be taken as the upper estimate Q^+ of the error probability Q of the restoring neuron

$$Q^{+} = \prod_{i=1}^{n+1} \left[(1-q_i) S^{a_i} + q_i S^{-a_i} \right].$$

The latter relation is easily rewritten in the equivalent form $rel = \frac{1}{2}$

$$Q^{+} = \prod_{i=1}^{n+1} Q_{i}^{+} = \prod_{i=1}^{n+1} \left[(1-q_{i})w_{i} + \frac{q_{i}}{w_{i}} \right],$$
(24)

where

$$\label{eq:Qi} Q_i^+ = (1-q_i) w_i + \frac{q_i}{w_i}$$
 and, along with this,

$$w_i = S^{a_i}, 0 < w_i < \infty, (i = \overline{1, n+1}).$$
 (25)

Now we can find the minimum Q_{\min}^+ of expression (24) and the value w_{0i} of w_i will attach a minimum to the upper estimate of the error probability Q^+ of the restoring neuron. For this, we use the conditions

$$\frac{\partial Q^+}{\partial w_i} = 0$$
$$i = \overline{1, n+1}$$

Hence

$$w_{0i} = \sqrt{\frac{q_i}{1 - q_i}}, \ (i = \overline{1, n + 1}).$$
 (26)

If (26) is substituted into expression (24), then by the second method for a minimal upper estimate of the error probability of the restoring neuron we obtain the relation

$$Q_{\min}^{+} = 2^{n+1} \cdot \prod_{i=1}^{n+1} \left[\sqrt{q_i (1-q_i)} \right], \tag{27}$$

which coincides with result (17) obtained by the first method. The weights $a_i(i = \overline{1, n+1})$ which match the error probabilities $q_i(i = \overline{1, n+1})$ are defined from relations (26) with notation (25) taken into account:

$$a_i = \frac{1}{2\ln S} \cdot \ln \frac{q_i}{1 - q_i} \\ i = \overline{1, n+1}$$

Since the value *S* satisfies condition (22), we have $\ln S < 0$ and therefore

$$\begin{array}{c} a_i = K \cdot \ln \frac{1 - q_i}{q_i} \\ i = \overline{1, n+1} \end{array} \right\},$$

$$(28)$$

where

(22)

$$K = \frac{1}{2|\ln S|}$$

$$0 < K < \infty$$
(29)

Thus, the weights $a_i(i = \overline{1, n+1})$, which are consistent with the error probabilities $q_i(i = \overline{1, n+1})$ and attach a minimum to the upper estimate of the error probability of the restoring neuron, are defined to within the a general positive factor *K*.

IV. CONCLUSION

A minimal upper estimate of the error probability of the restoring formal neuron is defined by formula (17) or, which is the same, by formula (27). In both cases the result can be written in the form

$$Q_{\min}^{+} = \exp\left(-\sum_{i=1}^{n+1} A(q_i)\right),$$
 (30)

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where

$$A(q_i) = \left| \ln \left[2\sqrt{q_i(1-q_i)} \right] \right|. \tag{31}$$

In view of relations (31) confirming the non-negativity of the values $A(q_i)$, formula (30) implies that an increase of the number *n* of inputs of the formal decision neuron bings about a monotone decrease of the minimal upper estimate of the error probability of restoration of the binary signal Q_{\min}^+ by the exponential law if, certainly, the error probabilities $q_i (i = \overline{1, n+1})$ at these inputs are not equal to $\frac{1}{2}$ when the minimal upper estimate of the error probability Q_{\min}^+ is equal to 1.

This result is demonstrates an essenatial inner connection with Shannon's theorem [10]. According to this theorem, the number of messages of length n (duration τ) composed of individual symbols – both in the absence and in the presence of fixed and probabilistic constraints (in the latter case it is assumed that the source is ergodic) – grows by the asymptotically exponential law as n (or τ) increases. In particular we understand this connection as follows: as the number n of inputs of the restoring formal neuron increases, the initial information to be used in making the decision Yincreases by the exponential law if there are a number of possible versions of the input signal, while the minimal upper estimate Q_{\min}^+ of the probability Q that the made decision is erroneous decreases by the same exponential law.

REFERENCES

[1] J. Neumann, "Probabilistic logics and synthesis of reliable organisms from unreliable components", in *Automata Studies: Annals of Mathematics Studies*, no. 34 (in Russian, translated from English), Moscow: Publishing House for foreign literature IIL, 1956, pp. 68-139.

[2] I. Varshavsky "Functional Possibilities and Synthesis of Threshold Elements" (in Russian), *Dokl. AN SSSR*, vol. 139, no. 5, pp. 1071-1074, 1961.

[3] M. Dertouzos *Threshold logic* (in Russian, translated from English), Moscow: Mir, 1966.

[4] C. R. Lageweg, S. D. Cotofana, S. Vassiliadis, "A full adder implementation using set based linear threshold gates", in *Proc. 9th IEEE International Conference on Electronics, Circuits and Systems*, New York, 2002, pp. 665–669.

[5] V. Beiu, J. M. Quintana, M. J. Avedillo, "VLSI Implementation of threshold Logic: A Comprehensive Survey", *IEEE Trans. Neural Networks*, vol. 14, pp. 1217–1243, May 2003.

[6] R. Zhang, P. Gupta, L. Zhong, N. K. Jha, "Threshold Network Synthesis and Optimization and Its Application to Nanotechnologies", *IEEE Transaction On Computer-Aided Design Of Integrated Circuits And Systems*, vol. 24, pp. 107–118, January 2005.

[7] T. Gowda, S. Vrudhula, N. Kulkarni, K. Berezowski, "Identification of Threshold Functions and Synthesis of Threshold Networks", *IEEE Transactions on Computer-Aided Design Of Integrated Circuits And Systems*, vol. 30, pp. 665-677, May 2011.

[8] W. Pierce, *Failure-tolerant computer design* (in Russian, translated from English), Moscow: Mir, 1968.

[9] G. A. Korn, T. M. Korn, *Mathematical handbook for scientists and engineers* (in Russian, translated from English), Moscow: Nauka, 1977, p. 832

[10] S. Goldman, *Information theory* (in Russian, translated from English), Moscow: Publishing House for foreign literature IIL, 1957.



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